

Radial Basis Function Neural Networks for Formation Control of Unmanned Aerial Vehicles

Duy-Nam Bui¹ and Manh Duong Phung*²

¹Vietnam National University, Hanoi, Vietnam.

²Fulbright University Vietnam, Ho Chi Minh City, Vietnam.

*Corresponding author. E-mail: duong.phung@fulbright.edu.vn.

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Abstract

This paper addresses the problem of controlling multiple unmanned aerial vehicles (UAVs) cooperating in a formation to carry out a complex task such as surface inspection. We first use the virtual leader-follower model to determine the topology and trajectory of the formation. A double-loop control system combining backstepping and sliding mode control techniques is then designed for the UAVs to track the trajectory. A radial basis function neural network (RBFNN) capable of estimating external disturbances is developed to enhance the robustness of the controller. The stability of the controller is proven by using the Lyapunov theorem. A number of comparisons and software-in-the-loop (SIL) tests have been conducted to evaluate the performance of the proposed controller. The results show that our controller not only outperforms other state-of-the-art controllers but is also sufficient for complex tasks of UAVs such as collecting surface data for inspection. The source code of our controller can be found at https://github.com/duynamrcv/rbf_bsmc.

1. Introduction

Unmanned aerial vehicles, when combined with computer vision technologies, can collect visual data of structures to provide valuable information for various tasks such as inspecting structural surfaces [1, 2], reconstructing 3D models [3, 4], identifying cracks [5, 6], and detecting corrosion and rust on steel bridges [7, 8]. However, using a single UAV for these tasks is inefficient due to the large size of the structures and the limited battery capacity of the UAV. A group of UAVs flying in a formation can be used to overcome those limitations [9–12]. The formation allows the UAVs to perform collaborative inspection to increase the efficiency and accuracy of data collection. It also allows for safe operation, as the formation control can prevent collision among the UAVs.

In formation control, the leader-follower approach is commonly used to provide flexibility in topology and trajectory selection [13, 14]. In the standard leader-follower method, one UAV is assigned as the leader, and the others are followers. The leader plays the role of a reference node for the followers to determine their locations to form the desired topology. The limitation of this approach, however, is the dependence of the system on the leader. If the leader is malfunctioning, the whole system will fail. The virtual leader-follower model can be used to cope with this problem. In this approach, the leader is purely a virtual entity, serving as a reference point for the followers to determine their positions [15]. By decoupling the physical leader from the model, this method mitigates the risk of complete system failure.

In the leader-follower model, linear controllers are commonly used to control individual UAVs to form the desired topology [15–17]. In [16], linear quadratic and neural networks-based controllers are combined to control a group of UAVs considering their full dynamics. In [17], the receding horizon control is employed to yield a fast convergence rate of the formation tracking control. This controller

also considers the orientation between the leader and the followers for accurate formation. The decentralized H_∞ -PID controller is introduced in [18] to maneuver a group of UAVs to deal with the external disturbance and trailing vortex coupling from their neighbor UAVs. A leader-follower formation control technique is presented in [19] to address issues related to backward error and suboptimal dynamic speed tracking in PID neural network control. Linear controllers, however, have limitations in handling constraints and parameter variation, especially when applied to nonlinear systems like UAVs.

In another approach, nonlinear controllers have been used for formation control [20–22]. In [23], first and second-order sliding-mode controllers are deployed to assure the asymptotic stability of the formation, taking into account modeling uncertainties. In [24], an adaptive controller using the dynamic estimation of the distance between the leader and the followers is introduced to address uncertainties related to positioning errors. Two finite-time observers are used in [25] to deal with bounded external disturbance force and torque. In [26], a non-uniform vector field that dynamically varies in magnitude and direction is employed to deal with the influence of wind in UAV formation control. A distributed model predictive control algorithm is introduced in [27] to coordinate the operation of a fleet of UAVs considering their spatial kinematics and unidirectional data transmissions. However, the convergence of these controllers depends on the characteristics of disturbances, which are hard to model due to their varying nature. A sufficient approach would be utilizing neural networks such as the radial basis function neural network (RBFNN) to estimate disturbances and use it as the feedback for control [28–31].

In this work, we present a new controller for a group of UAVs cooperating in a formation. The UAVs use the virtual leader-follower model to determine their trajectory and form the desired topology. The controller is developed using the backstepping and sliding mode control techniques. An RBFNN is then introduced to estimate external disturbances for better control performance. Our contributions in this work are as follows:

- i. The proposal of a new controller for UAV formation that is constructed by combining backstepping and sliding mode control techniques, thereby enabling the elimination of nonlinear components and enhancing system robustness. Additionally, the adverse effects associated with these controllers, such as “explosion of term” and “chattering”, are mitigated through the approximation of unknown factors by the neural network. As the result, the developed controller not only addresses the drawbacks of the aforementioned techniques but also augments the adaptability of the UAV system.
- ii. The design of a radial basis function neural network (RBFNN) that is capable of estimating external disturbances to compensate for input force control signals, thereby enabling the controller to maintain the required control quality.
- iii. The derivation of the stability proof for the designed controller using Lyapunov’s theorem, which is essential to ensure stable operation of the UAVs under conditions affected by external forces.
- iv. The comparison of the proposed controller with other popular methods including model predictive control (MPC), backstepping sliding mode control (BSMC) and sliding mode control (SMC) in different scenarios to confirm its superior performance. Software-in-the-loop tests were also conducted with a cooperative bridge inspection task to verify the validity of the proposed method for practical applications.

The rest of this paper is structured as follows. Section 2 presents the dynamic and formation models of the UAVs. Section 3 introduces the proposed controller. Section 4 shows evaluation results. The paper ends with conclusions described in 5.

2. Problem formulation

To control a group of UAVs, we first consider their dynamic model and formation topology with details as follows.

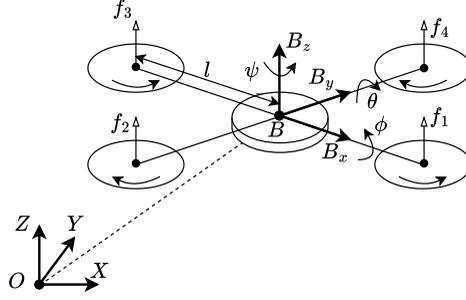


Figure 1: The structure of the quadrotor UAV in the global frame

2.1. UAV dynamic model

Consider a group of n UAVs, each is a quadrotor with two pairs of propellers rotating in opposite directions, as described in Figure 1. Frames $BB_xB_yB_z$ and $OXYZ$ are respectively the body-fixed and inertial frames. We use Euler angles to represent the attitude of the UAV. The configuration of the UAV includes its position $\xi = [x, y, z]^T$ and Euler angles $\Xi = [\phi, \theta, \psi]^T$, with $|\phi| \leq \pi/2$, $|\theta| \leq \pi/2$ and $|\psi| \leq \pi$. Those angles represent the roll, pitch and yaw orientation of the UAV, respectively. Control signals of the UAV are defined as follows:

$$\begin{bmatrix} f_t \\ \tau_\phi \\ \tau_\theta \\ \tau_\psi \end{bmatrix} = \begin{bmatrix} f_1 + f_2 + f_3 + f_4 \\ l(f_4 - f_2) \\ l(f_3 - f_1) \\ \tau_2 + \tau_4 - \tau_1 - \tau_3 \end{bmatrix}, \quad (1)$$

where l is the arm length; f_t is the total thrust of four propellers; $\tau_\phi, \tau_\theta, \tau_\psi$ are the torques in three axes; and f_i and τ_i , with $i = \{1, 2, 3, 4\}$, are the forces and torques generated by four propellers, respectively. According to [32], the dynamic model of the UAV is described as follows:

$$\begin{aligned} \ddot{x} &= (\cos \phi \sin \theta \cos \psi + \sin \phi \sin \psi) \frac{f_t}{m} + \frac{d_x}{m} \\ \ddot{y} &= (\cos \phi \sin \theta \sin \psi - \sin \phi \cos \psi) \frac{f_t}{m} + \frac{d_y}{m} \\ \ddot{z} &= \cos \phi \cos \theta \frac{f_t}{m} - g + \frac{d_z}{m} \\ \ddot{\phi} &= \frac{\dot{\theta} \dot{\psi} (I_y - I_z) + \tau_\phi}{I_x} \\ \ddot{\theta} &= \frac{\dot{\phi} \dot{\psi} (I_z - I_x) + \tau_\theta}{I_y} \\ \ddot{\psi} &= \frac{\dot{\phi} \dot{\theta} (I_x - I_y) + \tau_\psi}{I_z} \end{aligned} \quad (2)$$

where I_x, I_y, I_z are the moments of inertia, m is the mass of the UAV, g is the gravitational acceleration, and $[d_x, d_y, d_z]^T$ is the disturbance caused by factors such as wind or turbulent flows.

2.2. UAV formation model

The formation model used in this work is the virtual leader-follower model with two main components:

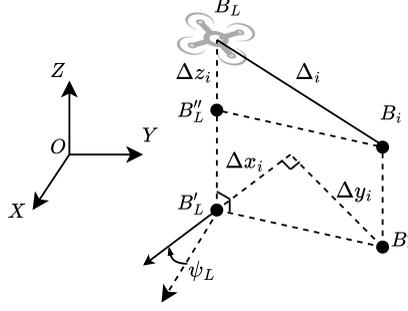


Figure 2: Illustration of the virtual leader-follower formation structure

- **Virtual leader:** a virtual leader is a non-physical UAV used as a reference for other UAVs to determine their position. Its trajectory represents the trajectory of the UAV group.
- **Follower:** a follower is a UAV that adjusts its position based on the virtual leader. Given the reference trajectory of the leader and the expected topology, the followers calculate their trajectories and then track them to form the desired formation.

Consider virtual leader B_L having position $\xi_L = [x_L, y_L, z_L]^T$ and heading angle ψ_L and follower B_i having position $\xi_i = [x_i, y_i, z_i]^T$ and yaw angle ψ_i . Let B'_L and B'_i be their projection on the OXY plane, respectively, B''_L be the projection of B_i on $B_L B'_L$, and $\Delta_i = [\Delta x_i, \Delta y_i, \Delta z_i]^T$ be the desired distance between follower B_i and the virtual leader, as depicted in Figure 2. Since $\Delta z_i = B_L B''_L$, the desired position of follower B_i can be computed as:

$$\begin{aligned} {}^d \xi_i &= \text{Rot}_z(\psi_L) \Delta_i + \xi_L \\ {}^d \psi_i &= \psi_L, \end{aligned} \quad (3)$$

where $\text{Rot}_z(\cdot) \in \mathbb{R}^{3 \times 3}$ is the rotation matrix around z-axis. Equation (3) allows the followers to compute their trajectory based on the trajectory of the virtual leader and the desired formation topology.

3. Controller design for UAV formation

Given the trajectory of the virtual leader, denoted as (ξ_L, ψ_L) , the desired trajectory of follower i , $({}^d \xi_i, {}^d \psi_i)$, in the formation can be computed based on (3). To track this trajectory, we design a dual-loop control system for each follower as shown in Figure 3. The outer loop is a position controller that regulates the altitude and horizontal position, while the inner loop is a backstepping sliding mode controller (BSMC) that handles the UAV's attitude, including its roll, pitch, and yaw angles. To account for external disturbances, the position controller was designed with a radial basis function neural network (RBFNN). A converter block is also included to convert the desired translational control forces into roll and pitch angles. Details of each controller are described as follows.

3.1. Position controller design

The position controller aims to keep the UAV's position aligned with the desired trajectory. It is designed based on BSMC with the use of RBFNN for disturbance estimation. According to (2), dynamic equations

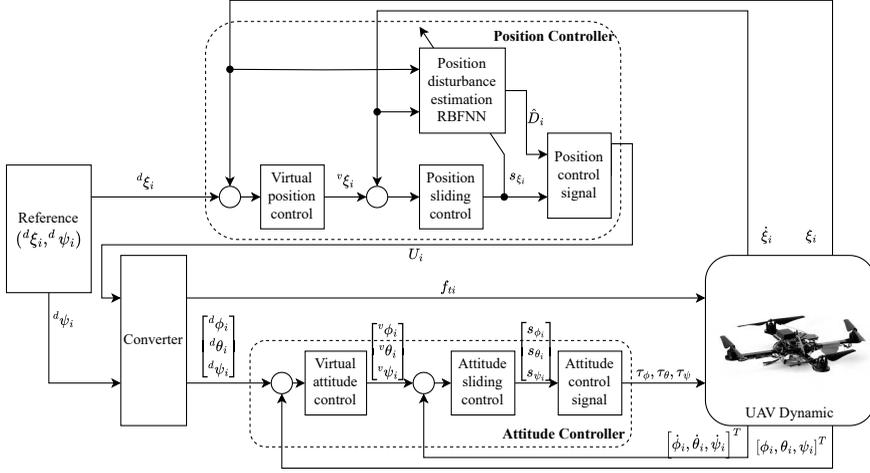


Figure 3: The proposed controller

for the translational motion of UAV i are given as follows:

$$\begin{aligned}\ddot{x}_i &= u_{xi} + \frac{d_{xi}}{m}, \\ \ddot{y}_i &= u_{yi} + \frac{d_{yi}}{m}, \\ \ddot{z}_i &= u_{zi} + \frac{d_{zi}}{m},\end{aligned}\quad (4)$$

where

$$\begin{aligned}u_{xi} &= (\cos \phi_i \sin \theta_i \cos \psi_i + \sin \phi_i \sin \psi_i) \frac{f_{ti}}{m}, \\ u_{yi} &= (\cos \phi_i \sin \theta_i \sin \psi_i - \sin \phi_i \cos \psi_i) \frac{f_{ti}}{m}, \\ u_{zi} &= \cos \phi_i \cos \theta_i \frac{f_{ti}}{m} - g.\end{aligned}\quad (5)$$

Let $\xi_i = [x_i, y_i, z_i]^T$ and $\dot{\xi}_i = [\dot{x}_i, \dot{y}_i, \dot{z}_i]^T$ respectively be the position and velocity of the translational motion, $U_i = [u_{xi}, u_{yi}, u_{zi}]^T$ be the control signal, and $D_i = \left[\frac{d_{xi}}{m}, \frac{d_{yi}}{m}, \frac{d_{zi}}{m} \right]^T$, with $\|D_i\| \leq \bar{d}$, be the external disturbance affecting UAV i . Equation (4) can be rewritten as:

$$\ddot{\xi}_i = U_i + D_i. \quad (6)$$

The BSMC is then designed as follows.

3.1.1. Backstepping sliding mode controller (BSMC) design

Let $e \xi_i$ be the translational error, $e \xi_i = \xi_i - d \xi_i$. The virtual velocity, $v \xi_i$, of the subsystem is designed as:

$$v \xi_i = d \dot{\xi}_i - \lambda_{\xi} e \xi_i, \quad (7)$$

where $\lambda_\xi > 0$ is a positive definite gain. The first candidate Lyapunov function is chosen as

$${}^1V_{\xi_i} = \frac{1}{2} e^{\xi_i^T} e^{\xi_i}. \quad (8)$$

Its derivative is given by

$${}^1\dot{V}_{\xi_i} = e^{\xi_i^T} e^{\dot{\xi}_i} = e^{\xi_i^T} (\dot{\xi}_i - d\dot{\xi}_i). \quad (9)$$

Substituting $\dot{\xi}_i = {}^v\dot{\xi}_i$ into (9) gives

$${}^1\dot{V}_{\xi_i} = -\lambda_\xi \|e^{\xi_i}\|^2 \leq 0. \quad (10)$$

Hence, the system is stable with the virtual velocity chosen in (7). The sliding mode control (SMC) algorithm is then utilized to design the input control signal for the position system. The sliding surface is chosen as follows:

$$s_{\xi_i} = \gamma_\xi e^{\xi_i} + (\dot{\xi}_i - {}^v\dot{\xi}_i), \quad (11)$$

where $\gamma_\xi > 0$ is a positive definite gain. Denote \hat{D}_i as the disturbance estimated via an estimator such as the RBFNN in Section 3.1.2. The derivative of s_{ξ_i} then can be obtained by using \hat{D}_i instead of D_i as follows:

$$\dot{s}_{\xi_i} = \gamma_\xi (\dot{\xi}_i - d\dot{\xi}_i) + U_i + \hat{D}_i - {}^v\dot{\xi}_i. \quad (12)$$

The second Lyapunov function of the subsystem is chosen as follows:

$${}^2V_{\xi_i} = \frac{1}{2} s_{\xi_i}^T s_{\xi_i} \quad (13)$$

The control signals are designed as follows:

$$\begin{aligned} U_{ieq} &= {}^v\dot{\xi}_i - \gamma_\xi (\dot{\xi}_i - d\dot{\xi}_i) - \hat{D}_i \\ U_{isw} &= -(c_{\xi 1} \text{sg}(s_{\xi_i}) + c_{\xi 2} s_{\xi_i}), \end{aligned} \quad (14)$$

where $c_{\xi 1}$ and $c_{\xi 2}$ are positive gains, U_{ieq} is the equivalent control signal that maintains the position variables on the sliding manifold, U_{isw} is the signal that leads the subsystem to the sliding surface s_{ξ_i} , and $\text{sg}(\cdot)$ is the piece-wise continuous function defined as

$$\text{sg}(x) = \begin{cases} 1 & x > \epsilon \\ -1 & x < -\epsilon \\ \frac{x}{\epsilon} & \text{otherwise} \end{cases} \quad (15)$$

where $0 < \epsilon < 1$ is a pre-defined constant.

Theorem 3.1. *Consider the position control system of the UAV. If the control signal is chosen as*

$$U_i = U_{ieq} + U_{isw}, \quad (16)$$

the system is stable.

Proof. Taking the first derivative of ${}^2V_{\xi_i}$ gives

$${}^2\dot{V}_{\xi_i} = s_{\xi_i}^T \dot{s}_{\xi_i}. \quad (17)$$

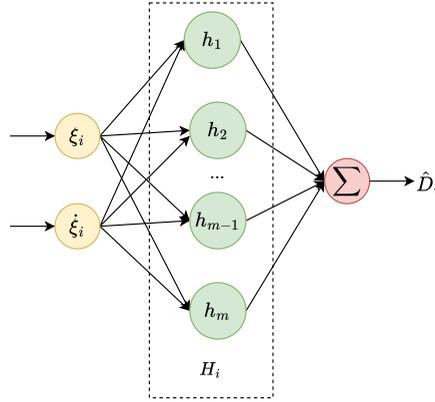


Figure 4: The RBFNN structure

Substituting (12) into (17) gives

$${}^2\dot{V}_{\xi_i} = s_{\xi_i}^T \left(\gamma_{\xi} \left(\dot{\xi}_i - d \dot{\xi}_i \right) + U_i + \hat{D}_i - v \dot{\xi}_i \right). \quad (18)$$

By substituting (14) and (16) into (18), \dot{V}_{ξ_i} becomes

$${}^2\dot{V}_{\xi_i} = -c_{\xi 1} s_{\xi_i}^T \text{sg} (s_{\xi_i}) - c_{\xi 2} \|s_{\xi_i}\|^2 \leq 0. \quad (19)$$

According to Lyapunov's stability theorem, the system is stable. \square

3.1.2. Radial basis function neural network (RBFNN) design

During operation, UAVs are subject to inevitable disturbances such as wind or turbulent flows. Those disturbances affect the system performance, but are complex to model and analyze. We address this problem by exploiting the online learning capability of neural networks to estimate the disturbances. Previous studies on universal approximation theorems for RBFNN show that RBFNN can approximate any nonlinear function on a bounded set with an arbitrary level of accuracy [31]. In this work, we design a disturbance estimator using a neural network with the radial basis function (RBF). The network has three layers including an input layer, a hidden layer, and an output layer, as shown in Figure 4. Position vector ξ_i and its derivation $\dot{\xi}_i$ are the input of the network. At the hidden layer, neurons are activated by a radial basis function. The output of neuron j is computed as:

$$h_j = \exp \left(- \frac{\|\xi_i - \mu_{1j}\|^2 + \|\dot{\xi}_i - \mu_{2j}\|^2}{b^2} \right), \quad (20)$$

where b is a parameter controlling the width of the Gaussian function, μ_{1j} and μ_{2j} are predefined center points, and $j \in \{1, 2, \dots, m\}$ is the neuron index with m being the number of neurons in the hidden layer. The output layer is a weighted sum. Let W_i be the optimal weight matrix, H_i be the output of the hidden layer, and σ_i be the approximation error. Disturbance D_i affecting UAV i then can be expressed by:

$$D_i = W_i^T H_i + \sigma_i \quad (21)$$

The output \hat{D}_i of the RBFNN approximates D_i as:

$$\hat{D}_i = \hat{W}_i^T H_i, \quad (22)$$

where \hat{W}_i is a trained weight matrix. This matrix is updated based on the following rule:

$$\dot{\hat{W}}_i = a \left(H_i s_{\xi_i}^T - \eta \|s_{\xi_i}\| \hat{W}_i \right), \quad (23)$$

where a is a positive definite gain matrix. With this structure, the estimation of disturbance D_i can be described as in Algorithm 1, where $\beta \in (0, 1)$ is the momentum factor.

Algorithm 1: Pseudocode to estimate disturbances by RBFNN for UAV i

```

/* Initialization only one time */
1 Initialize parameters  $m, a, b, \eta, \beta$ ;
2 Initialize predefined center points  $\mu_1, \mu_2$ ;
3 Create random weight matrix  $\hat{W}_i$ ;
4 Initialize stored weight matrices  $\hat{W}_{1i} = \hat{W}_{2i} = \hat{W}_i$ ;
/* Online training */
5 foreach computation step do
6   Get the current values of  $\xi_i, \dot{\xi}_i, s_{\xi_i}$ ;
7    $H_i = \text{zeros}(m, 1)$ ;
8   for  $j = 1$  to  $m$  do
9     Compute sub-hidden layer  $h_j$ ; /* Equation 20 */
10    Compute the value of  $\hat{W}_i$ ; /* Equation 23 */
11    Update the weight matrix  $\hat{W}_i = \hat{W}_{1i} + \dot{\hat{W}}_i + \beta(\hat{W}_{1i} - \hat{W}_{2i})$ ;
12    Update stored weight matrices  $\hat{W}_{2i} = \hat{W}_{1i}$  and  $\hat{W}_{1i} = \hat{W}_i$ ;
13    Compute the current estimated disturbance  $\hat{D}_i$ ; /* Equation 22 */

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3.1.3. Stability of the position controller

The stability of this controller is addressed in Theorem 3.2 as follows.

Theorem 3.2. Consider UAV i affected by external disturbance D_i as described in (6), the control signals designed in (14), the bounded weight $\|W_i\| \leq \bar{W}_i$, and the update rule for RBFNN in (23). If the following inequality condition is satisfied:

$$\|s_{\xi_i}\| \geq \frac{\bar{\sigma} + \frac{1}{4}\eta\bar{W}_i^2}{c\xi_2}, \quad (24)$$

the position control system is stable.

Proof. Choose the candidate Lyapunov function as follows:

$$V_{\xi_i} = {}^2V_{\xi_i} + \frac{1}{2}\text{tr} \left(\tilde{W}_i^T a^{-1} \tilde{W}_i \right), \quad (25)$$

where $\tilde{W}_i = W_i - \hat{W}_i$ is the error weight matrix. Taking the first derivative of V_{ξ_i} gives:

$$\begin{aligned}
\dot{V}_{\xi_i} &= s_{\xi_i}^T \dot{s}_{\xi_i} + \text{tr} \left(\tilde{W}_i^T a^{-1} \dot{\tilde{W}}_i \right) \\
&= -c_{\xi 1} s_{\xi_i}^T \text{sg} (s_{\xi_i}) - c_{\xi 2} \|s_{\xi_i}\|^2 - s_{\xi_i}^T (\hat{D}_i - D_i) - \text{tr} \left(\tilde{W}_i^T a^{-1} \dot{\tilde{W}}_i \right) \\
&= -c_{\xi 1} s_{\xi_i}^T \text{sg} (s_{\xi_i}) - c_{\xi 2} \|s_{\xi_i}\|^2 + s_{\xi_i}^T \sigma_i + s_{\xi_i}^T \tilde{W}_i^T H_i - \text{tr} \left(\tilde{W}_i^T a^{-1} \dot{\tilde{W}}_i \right) \\
&= -c_{\xi 1} s_{\xi_i}^T \text{sg} (s_{\xi_i}) - c_{\xi 2} \|s_{\xi_i}\|^2 + s_{\xi_i}^T \sigma_i + \text{tr} \left(-\tilde{W}_i^T \left(a^{-1} \dot{\tilde{W}}_i - H_i s_{\xi_i}^T \right) \right)
\end{aligned} \tag{26}$$

With the updated rule of the neural network, \dot{V}_{ξ_i} can be rewritten as follows:

$$\begin{aligned}
\dot{V}_{\xi_i} &= -c_{\xi 1} s_{\xi_i}^T \text{sg} (s_{\xi_i}) - c_{\xi 2} \|s_{\xi_i}\|^2 + s_{\xi_i}^T \sigma_i + \text{tr} \left(\tilde{W}_i^T \eta \|s_{\xi_i}\| \hat{w}_i \right) \\
&= -c_{\xi 1} s_{\xi_i}^T \text{sg} (s_{\xi_i}) - c_{\xi 2} \|s_{\xi_i}\|^2 + s_{\xi_i}^T \sigma_i + \eta \|s_{\xi_i}\| \text{tr} \left(\tilde{W}_i^T (W_i - \tilde{W}_i) \right)
\end{aligned} \tag{27}$$

According to the Cauchy-Schwarz inequality, the following inequality equation can be satisfied:

$$\text{tr} \left(\tilde{W}_i^T (W_i - \tilde{W}_i) \right) \leq \|\tilde{W}_i\| \|W_i\| - \|\tilde{W}_i\|^2 \tag{28}$$

Thus,

$$\begin{aligned}
\dot{V}_{\xi_i} &\leq -c_{\xi 1} s_{\xi_i}^T \text{sg} (s_{\xi_i}) - c_{\xi 2} \|s_{\xi_i}\|^2 + \|s_{\xi_i}\| \bar{\sigma} + \eta \|s_{\xi_i}\| \left(\|\tilde{W}_i\| \|W_i\| - \|\tilde{W}_i\|^2 \right) \\
&\leq -c_{\xi 1} s_{\xi_i}^T \text{sg} (s_{\xi_i}) - c_{\xi 2} \|s_{\xi_i}\|^2 + \|s_{\xi_i}\| \bar{\sigma} + \frac{1}{4} \eta \|s_{\xi_i}\| \bar{W}_i^2 - \eta \|s_{\xi_i}\| \left(\frac{1}{2} \|W_i\| - \|\tilde{W}_i\| \right)^2
\end{aligned} \tag{29}$$

Based on an extension of the Lyapunov theorem [33], $\|s_{\xi_i}\|$ is bounded. Moreover, the control gain $c_{\xi 2}$ can be selected large enough so that

$$\left[\bar{\sigma} + \eta \bar{W}_i^2 / 4 \right] / c_{\xi 2} \leq b_{\xi} \tag{30}$$

Therefore, with the inequality condition (24), \dot{V}_p can be rewritten as follows:

$$\dot{V}_{\xi_i} \leq -c_{\xi 1} s_{\xi_i}^T \text{sg} (s_{\xi_i}) - \eta \|s_{\xi_i}\| \left(\frac{1}{2} \|W_i\| - \|\tilde{W}_i\| \right)^2 \leq 0 \tag{31}$$

The Lyapunov stability condition is satisfied. \square

3.2. Attitude controller design

In our system, the position controller is the outer loop of the UAV control system, as depicted in Figure 3. Its control signal is then fed to the converter block to calculate the desired angles and translational

forces based on (5) as:

$$\begin{aligned} d\theta_i &= \arctan\left(\frac{u_{xi} \cos^d \psi_i + u_{yi} \sin^d \psi_i}{u_{zi} + g}\right) \\ d\phi_i &= \arctan\left(\cos^d \theta_i \frac{u_{xi} \sin^d \psi_i - u_{yi} \cos^d \psi_i}{u_{zi} + g}\right) \\ f_{1i} &= \frac{u_{zi} + g}{\cos^d \phi_i \cos^d \theta_i} \end{aligned} \quad (32)$$

They are used as the reference for the attitude controller, which is designed based on the BSMC. From (2), the dynamic equation for the roll angle is given by:

$$\ddot{\phi}_i = \frac{\dot{\theta}_i \dot{\psi}_i (I_y - I_z) + \tau_{\phi_i}}{I_x}. \quad (33)$$

Denote ${}^e \phi_i = \phi_i - {}^d \phi_i$ as the roll angle error. The virtual velocity, ${}^v \phi_i$, is defined as:

$${}^v \phi_i = {}^d \dot{\phi}_i - \lambda_\phi \phi_{1e}, \quad (34)$$

where $\lambda_\phi > 0$ is a positive gain. The first candidate Lyapunov function for subsystem ϕ_{1e} is chosen as:

$${}^1 V_{\phi_i} = \frac{1}{2} {}^e \phi_i^2. \quad (35)$$

Taking the first derivative of ${}^1 V_{\phi_i}$ gives:

$${}^1 \dot{V}_{\phi_i} = {}^e \phi_i {}^e \dot{\phi}_i = {}^e \phi_i (\dot{\phi}_i - {}^d \dot{\phi}_i). \quad (36)$$

Substituting $\dot{\phi}_i = {}^v \phi_i$ into (36) gives

$${}^1 \dot{V}_{\phi_i} = -\lambda_\phi {}^e \phi_i^2 \leq 0. \quad (37)$$

Thus, the Lyapunov stability is guaranteed. The sliding surface of the roll angle subsystem is expressed as:

$$s_{\phi_i} = \gamma_\phi {}^e \phi_i + (\dot{\phi}_i - {}^e \phi_i), \quad (38)$$

where $\gamma_\phi > 0$ is a positive gain. The first derivative of s_{ϕ_i} is given by:

$$\dot{s}_{\phi_i} = \gamma_\phi (\dot{\phi}_i - {}^d \dot{\phi}_i) + \frac{\dot{\theta}_i \dot{\psi}_i (I_y - I_z) + \tau_{\phi_i}}{I_x} - {}^v \dot{\phi}_i. \quad (39)$$

The control signal is then designed with two sub-control signals, ${}^{eq} \tau_{\phi_i}$ and ${}^{sw} \tau_{\phi_i}$. ${}^{eq} \tau_{\phi_i}$ is the equivalent control signal that maintains the roll angle on the sliding manifold and ${}^{sw} \tau_{\phi_i}$ is the signal that leads the subsystem to the sliding surface s_{ϕ_i} . They are chosen as follows:

$$\begin{aligned} {}^{eq} \tau_{\phi_i} &= I_x \left({}^v \dot{\phi}_i - \gamma_\phi (\dot{\phi}_i - {}^d \dot{\phi}_i) \right) - \dot{\theta}_i \dot{\psi}_i (I_y - I_z) \\ {}^{sw} \tau_{\phi_i} &= -I_x (c_{\phi 1} \text{sg}(s_{\phi_i}) + c_{\phi 2} s_{\phi_i}), \end{aligned} \quad (40)$$

where $c_{\phi 1}$ and $c_{\phi 2}$ are positive gains.

Theorem 3.3. Consider the roll angle subsystem (33). If the control signal is designed as:

$$\tau_{\phi_i} = {}^{eq}\tau_{\phi_i} + {}^{sw}\tau_{\phi_i}, \quad (41)$$

the roll angle control system is stable.

Proof. The candidate Lyapunov function of the roll angle subsystem is chosen as follows:

$$V_{\phi_i} = \frac{1}{2}s_{\phi_i}^2. \quad (42)$$

Taking the first derivative of V_{ϕ_i} gives

$$\dot{V}_{\phi_i} = s_{\phi_i}\dot{s}_{\phi_i}. \quad (43)$$

By substituting (39) into (43), we have

$$\dot{V}_{\phi_i} = s_{\phi_i} \left(\gamma_{\phi} (\dot{\phi}_i - {}^d\dot{\phi}_i) + \frac{\dot{\theta}_i \psi_i (I_y - I_z) + \tau_{\phi_i}}{I_x} - {}^v\dot{\phi}_i \right). \quad (44)$$

Finally, substituting (40) and (41) into (44) gives

$$\dot{V}_{\phi_i} = -c_{\phi 1} s_{\phi_i} \text{sg}(s_{\phi_i}) - c_{\phi 2} s_{\phi_i}^2 \leq 0. \quad (45)$$

Thus, the Lyapunov stability of the roll angle control system is guaranteed. \square

The control signals for the pitch and yaw angles can be obtained by applying the design process similar to the roll angle. As a result, the pitch control signals are obtained as:

$$\begin{aligned} \tau_{\theta_i} &= {}^{eq}\tau_{\theta_i} + {}^{sw}\tau_{\theta_i} \\ {}^{eq}\tau_{\theta_i} &= I_y \left({}^v\dot{\theta}_i - \gamma_{\theta} (\dot{\theta}_i - {}^d\dot{\theta}_i) \right) - \dot{\phi}_i \psi_i (I_z - I_x) \\ {}^{sw}\tau_{\theta_i} &= -I_y (c_{\theta 1} \text{sg}(s_{\theta_i}) + c_{\theta 2} s_{\theta_i}), \end{aligned} \quad (46)$$

and the yaw control signals are given by:

$$\begin{aligned} \tau_{\psi_i} &= {}^{eq}\tau_{\psi_i} + {}^{sw}\tau_{\psi_i} \\ {}^{eq}\tau_{\psi_i} &= I_z \left({}^v\dot{\psi}_i - \gamma_{\psi} (\dot{\psi}_i - {}^d\dot{\psi}_i) \right) - \dot{\phi}_i \dot{\theta}_i (I_x - I_y) \\ {}^{sw}\tau_{\psi_i} &= -I_z (c_{\psi 1} \text{sg}(s_{\psi_i}) + c_{\psi 2} s_{\psi_i}). \end{aligned} \quad (47)$$

4. Results

To evaluate the performance of the proposed control system, we have conducted a number of evaluations and comparisons¹. The UAV model used is the Hummingbird quadrotors [34], whose parameters are shown in Table 1. Parameters of the position and attitude controllers are chosen as shown in Table 2. The desired formation is a triangular shape with $\Delta_1 = [2, 0, 0]^T$, $\Delta_2 = [0, 0, 2]^T$, and $\Delta_3 = [0, 0, -2]^T$, as depicted in Figure 5. Comparisons are conducted between the proposed controller (RBF-BSMC) and three other controllers namely model predictive control (MPC) [27, 35], backstepping sliding mode control (BSMC) [36, 37] and sliding mode control (SMC) [23, 38] in different scenarios.

¹Evaluation results in Scenario 1 and Scenario 2 - <https://youtu.be/LYD7269n1-c>

Table 1: Parameters of the Hummingbird quadrotor

Description	Notation	Value
The mass of UAV (kg)	m	0.68
The gravitational acceleration (m/s ²)	g	9.81
Moment of inertia about x and y axes (kg.m ²)	I_x, I_y	0.007
Moment of inertia about z-axis (kg.m ²)	I_z	0.012
Length of UAV arm (m)	l	0.17
Thrust coefficient	k_t	29×10^{-6}
Drag coefficient	k_d	1.1×10^{-6}

Table 2: Parameters of the proposed controller

Parameters of the BSMC	
Attitude controller	$\lambda_\phi = \lambda_\theta = \lambda_\psi = 5;$ $\gamma_\phi = \gamma_\theta = \gamma_\psi = 5;$ $c_{\phi 1} = c_{\theta 1} = c_{\psi 1} = 2;$ $c_{\phi 2} = c_{\theta 2} = c_{\psi 2} = 2;$
Position controller	$\lambda_\xi = 3; \gamma_\xi = 2; c_{\xi 1} = c_{\xi 2} = 2$
Parameters of the RBFNN	
Position controller	$m = 50; a = 0.1; b = 10; \eta = 0.2;$ $\begin{cases} \mu_x = \text{linspace}(-r_x, r_x, n) \\ \mu_y = \text{linspace}(-r_y, r_y, n) \\ \mu_z = \text{linspace}(-r_z, r_z, n) \end{cases}; r_x = r_y = r_z = 15;$ $\mu_1 = [\mu_x, \mu_y, \mu_z]^T; \mu_2 = \mu_1;$

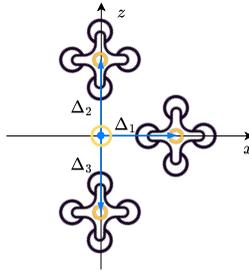


Figure 5: The desired topology

4.1. Scenario 1

In this scenario, external disturbances acting on the formation are generated based on the combination of the rectangle and full wavelength “1-cosine” wind model [39], as shown in Figure 6. The initial positions of the UAVs are set as $\xi_1 = [3, 2, 4]^T$, $\xi_2 = [2, 1, 4]^T$, and $\xi_3 = [0, 0, 4]^T$. The desired trajectory of the virtual leader is a spiral with the z coordinate increasing over time as expressed in (48).

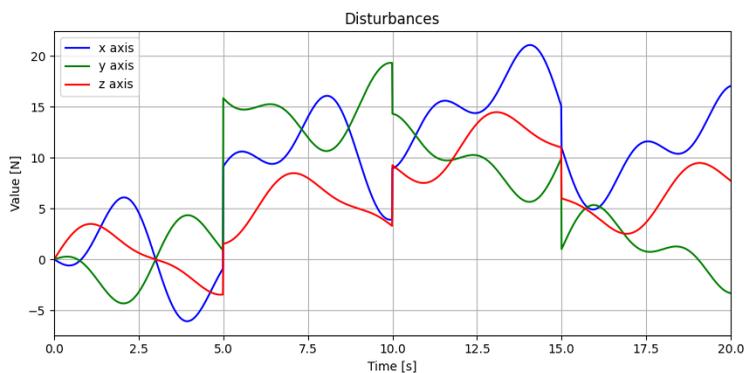


Figure 6: The external disturbance acting on the formation generated based on the rectangle and full wavelength “1-cosine” wind model in Scenario 1

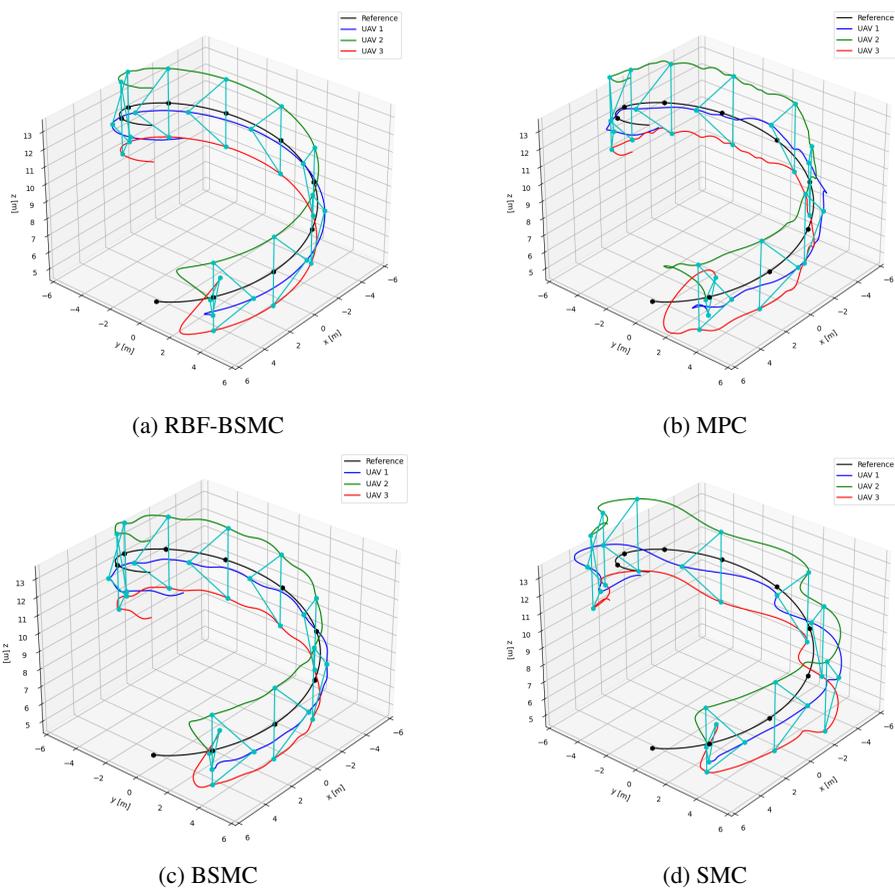


Figure 7: Trajectories of the UAV formation generated by the four controllers in Scenario 1

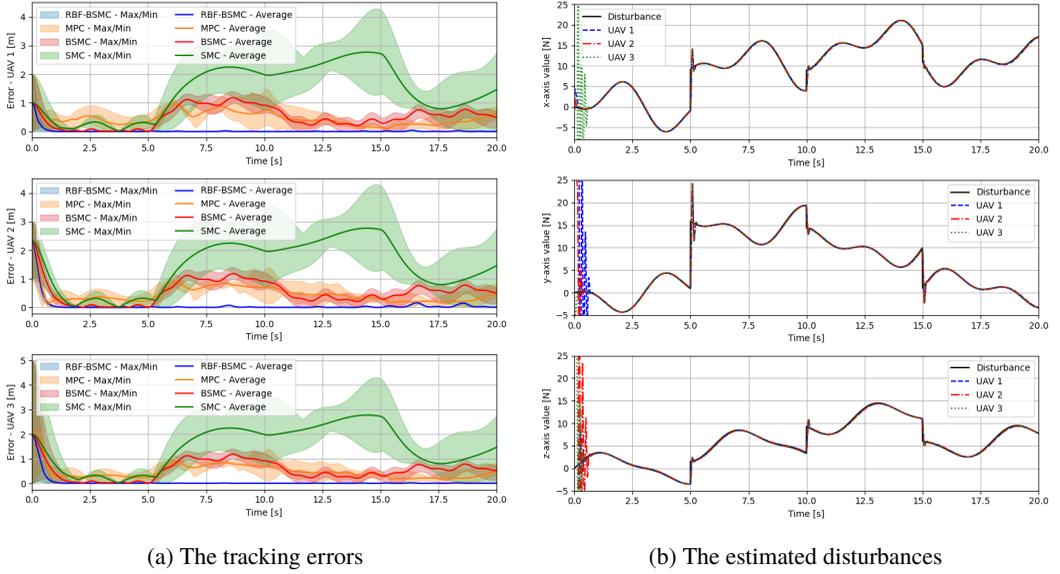


Figure 8: The tracking errors and estimated disturbances of the controllers in Scenario 1

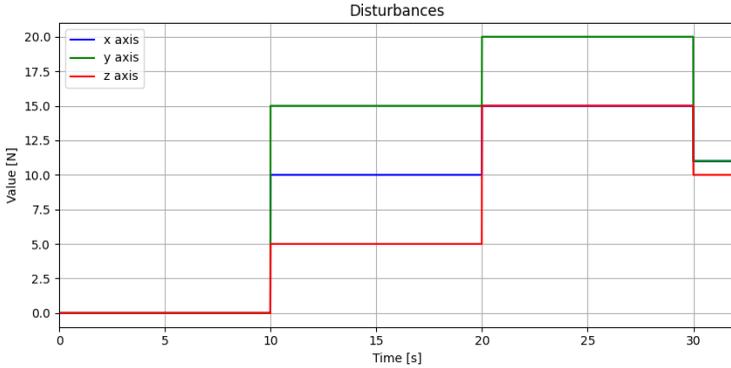


Figure 9: The external disturbance acting on the formation generated based on the rectangle wind model in Scenario 2

$$\begin{aligned}
 x_L &= 5 \cos\left(\frac{2\pi}{20}t\right) \text{ (m)}, \\
 y_L &= 5 \sin\left(\frac{2\pi}{20}t\right) \text{ (m)}, \\
 z_L &= 0.5t + 5 \text{ (m)}, \\
 \psi_L &= \frac{2\pi}{20}t + \frac{\pi}{2} \text{ (rad)}.
 \end{aligned} \tag{48}$$

Figure 7 shows the 3D views of the trajectory tracking results of the UAV formation. It can be seen that the UAVs quickly reach the initial positions to form the desired shape. They then maintain the shape

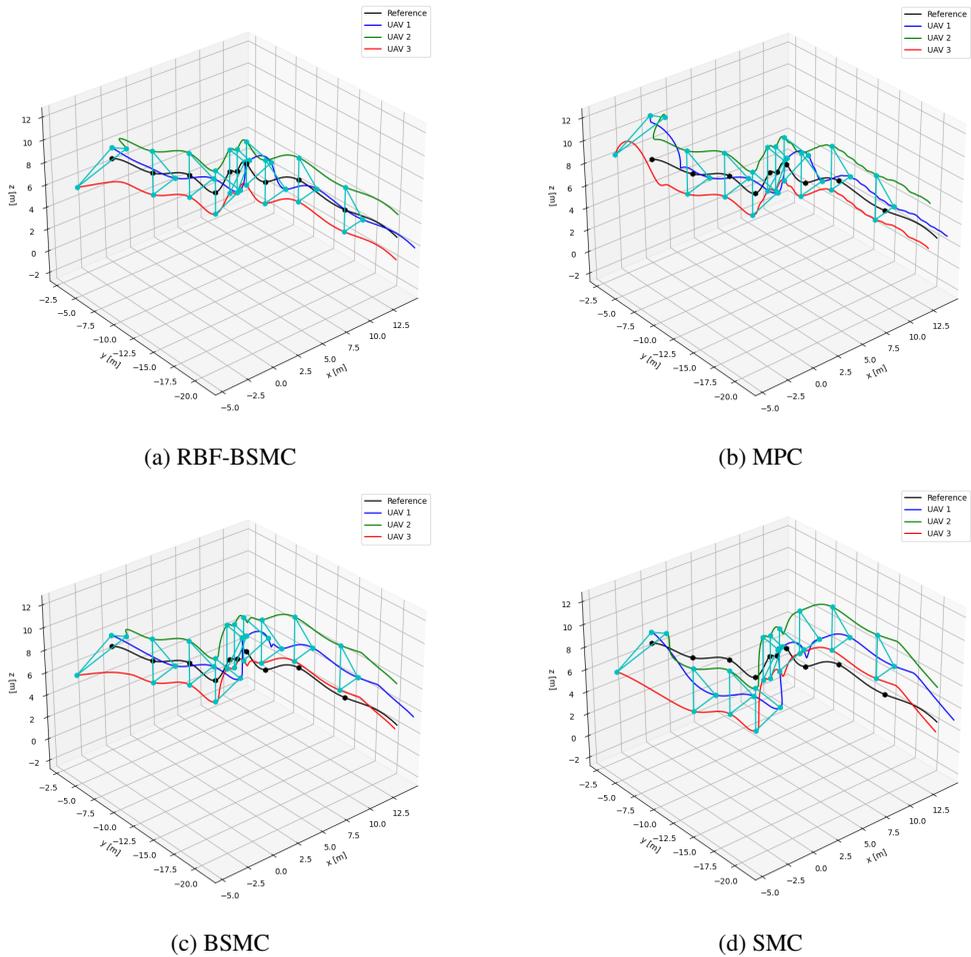


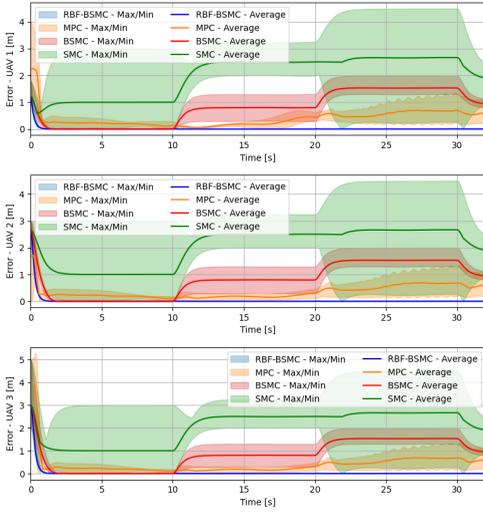
Figure 10: Trajectories of the UAV formation generated by the four controllers in Scenario 2

while following the reference trajectory. However, the trajectory of the proposed method is smoother and more accurate than the others due to its capability to estimate the disturbance via the RBFNN and use it as feedback to adjust the control signals. This result can be further verified via the tracking errors as shown in Figure 8a. It can be seen that the average tracking errors of the proposed controller quickly converge to zeros, whereas those errors of the other controllers largely fluctuate due to disturbances. In addition, the maximum and minimum tracking errors of the UAVs are also very small with our method, which confirm its stability for formation control.

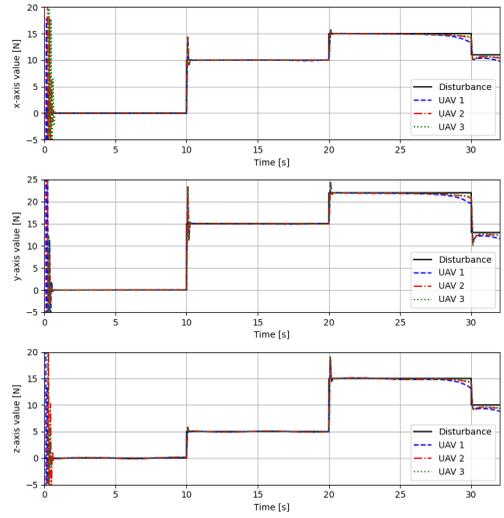
Figure 8b shows the disturbances estimated for each UAV by the proposed controller. After the transition period, the estimation starts to converge to the real disturbance. This provides feedback for the controller to adjust the control signal for better tracking performance.

4.2. Scenario 2

In this scenario, the rectangle wind model is used to generate external disturbances as shown in Figure 9. The initial positions of the UAVs are set as $\xi_1 = [2, 0, 9]^T$, $\xi_2 = [5, 2, 7]^T$, and $\xi_3 = [-3, -2, 8]^T$. The



(a) The tracking errors

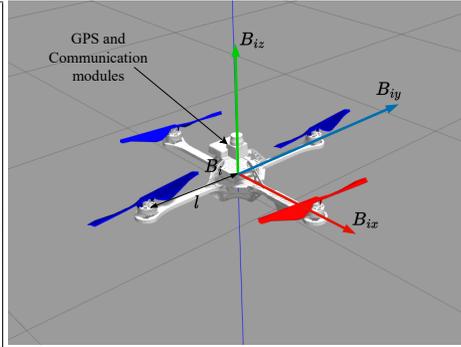


(b) The estimated disturbances

Figure 11: The tracking errors and estimated disturbances of the controllers in Scenario 2

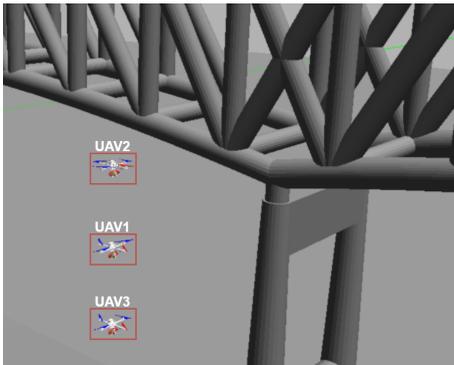


(a) The real Hummingbird drone

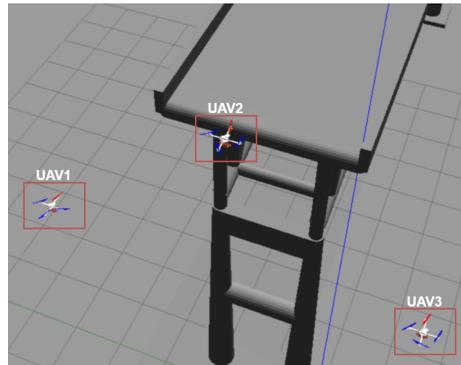


(b) The Hummingbird model

Figure 12: The Hummingbird drone model used in validation [32, 34]

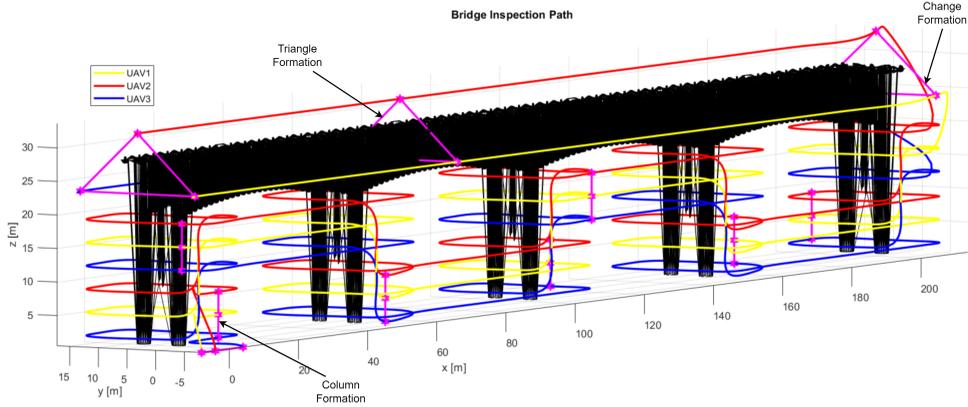


(a) Vertical formation

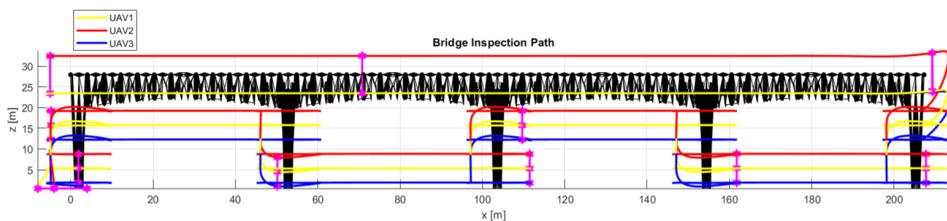


(b) Triangular formation

Figure 13: The vertical and triangular formation topologies used in SIL tests



(a) 3D view



(b) Side view



(c) Top view

Figure 14: The planned paths to inspect the bridge

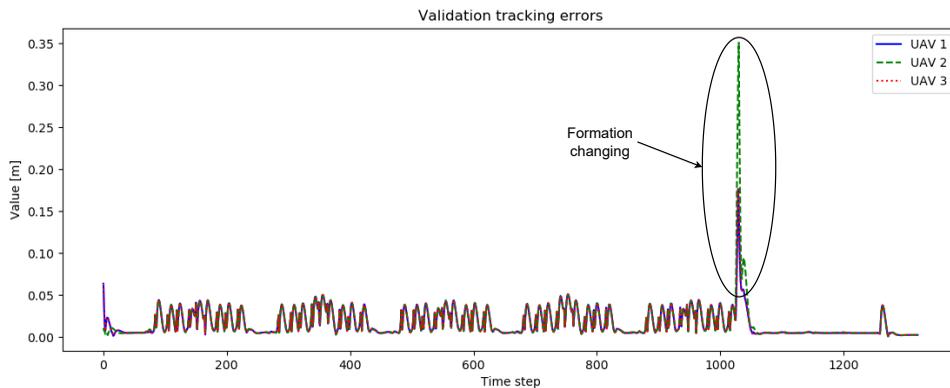


Figure 15: The tracking errors of UAV formation in the Gazebo SIL

desired trajectory of the virtual leader is generated based on an inspection path used to collect surface data [40].

The 3D of the UAV formation tracking results are shown in Figure 10. It can be seen that all controllers are able to drive the UAVs to reach their reference positions and then track them to form the desired shape. The proposed controller, however, introduces smaller tracking errors than the other controllers, as shown in Figure 11a, due to its disturbance estimator. As shown in Figure 11b, the estimation closely follows the actual disturbances except for the positions where step changes happen. At those positions, a transition period is needed for the estimator to converge to the new steady state. However, the settling time of the estimator is short allowing it to provide timely feedback to the controller.

4.3. Validation with software-in-the-loop tests

To further validate the proposed control system, we have carried out software-in-the-loop (SIL) tests that involve the inspection of a scaled-down 3D model of a real bridge with 5 columns, as shown in Figure 14. The UAV model used is a Hummingbird quadrotor² developed based on Gazebo-based RotorS simulator [32], as depicted in Figure 12. The formation used includes two topologies, vertical and triangular shapes, as shown in Figure 13. According to our previous work [41], the generated path to inspect the bridge includes two stages. The first stage covers all columns of the bridge using a vertical formation. The second one uses a triangular formation to cover the side and top surfaces of the bridge, as depicted in Figure 14.

Given the planned paths, the UAVs start to fly from positions $[-8, -8, 0]^T$, $[-4, -8, 0]^T$, $[4, -8, 0]^T$, and reach their initial positions to form a vertical formation as shown in Figure 14a. The formation then tracks the planned path to acquire surface images of the bridge³. Figure 15 shows the tracking errors of the UAVs during operation. It can be seen that the errors quickly converge to small values in both inspection stages except between time steps 1070 and 1120, where there is a change in the formation topology. The errors in the first stage are slightly larger than in the second one as the UAVs frequently changes their direction to navigate around each column of the bridge. Nevertheless, the average tracking error of less than 5 cm is sufficient for most UAV-related applications and thus confirms the validity of our approach.

5. Conclusion

In this paper, we have presented a robust control system using RBFNN for a group of UAVs flying in a formation. By combining BSC with SMC, the controller can handle nonlinearity to increase its control performance. The use of RBFNN enables the system to estimate external disturbances to enhance its control robustness. By using Lyapunov's theorem, we proved that the control system is stable and the proposed controller can track the reference trajectory. Evaluation results show that the proposed controller outperforms the state-of-the-art BSMC in terms of accuracy and robustness and is sufficient for most UAV applications.

Author Contributions. Duy-Nam Bui: Conceptualization, Methodology, Implementation, Writing - original draft. Manh Duong Phung: Investigation, Conceptualization, Supervision, Writing - review and editing.

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Conflicts of Interest. The authors declare no conflicts of interest exist.

²Source code used for Gazebo validation - https://github.com/duynamrcv/hummingbird_simulator

³SIL validation - <https://youtu.be/1yUCzWRDcp0>

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